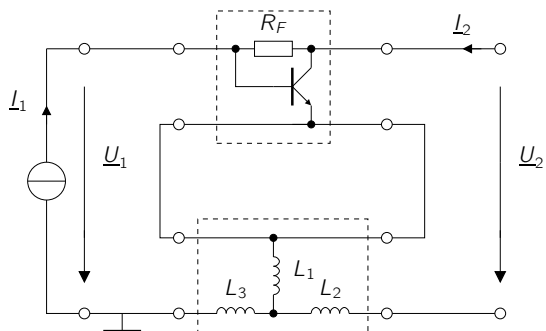
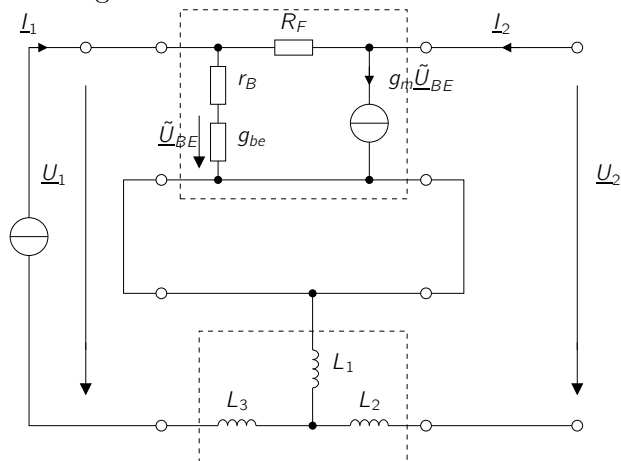


### Aufgabe 1

a) Wechselstromersatzschaltbild



b) Kleinsignalersatzschaltbild



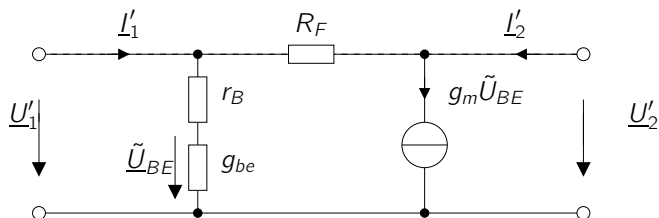
c) (i) seriell-seriell-Kopplung (SSK)

(ii) Z-Parameter, da

$$\begin{pmatrix} U_1 \\ U_2 \end{pmatrix} = \begin{pmatrix} U_1' \\ U_2' \end{pmatrix} + \begin{pmatrix} U_1'' \\ U_2'' \end{pmatrix} = [Z'] \begin{pmatrix} I_1' \\ I_2' \end{pmatrix} + [Z''] \begin{pmatrix} I_1'' \\ I_2'' \end{pmatrix} = \underbrace{([Z'] + [Z''])}_{[Z]} \begin{pmatrix} I_1 \\ I_2 \end{pmatrix}$$

⇒ Gesamtmatrix ist die Summe der Teilmatrizen

d) Hauptzweitor:



$$\begin{aligned} \underline{Z}'_{11} &= \left. \frac{U'_1}{I'_1} \right|_{I'_2=0} : I'_2 = \frac{U'_1}{r_B + \frac{1}{g_{be}}} + g_m \tilde{U}_{be} = \frac{U'_1}{r_B + \frac{1}{g_{be}}} + g_m \frac{1}{r_B + \frac{1}{g_{be}}} U'_1 \\ &= \frac{1 + \beta}{r_b + \frac{1}{g_{be}}} U'_1 \\ \Rightarrow \underline{Z}'_{11} &= \frac{r_B + \frac{1}{g_{be}}}{1 + \beta_0} = \frac{1 + r_B g_{be}}{g_{be} + g_m} \end{aligned}$$

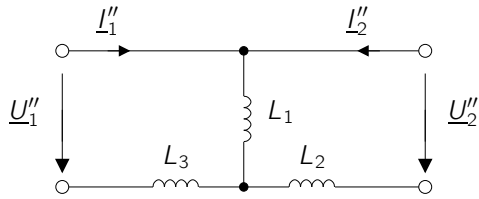
$$\begin{aligned} \underline{Z}'_{12} &= \left. \frac{U'_1}{I'_2} \right|_{I'_1=0} : I'_2 = g_m \tilde{U}_{BE} + g_{be} \tilde{U}_{BE} ; \tilde{U}_{BE} = U'_1 \frac{\frac{1}{g_{be}}}{r_B + \frac{1}{g_{be}}} = \frac{1}{1 + g_{be} r_B} U'_1 \\ \Rightarrow I'_2 &= (g_m + g_{be}) \frac{1}{1 + g_{be} r_B} U'_1 \\ \Rightarrow \underline{Z}'_{12} &= \frac{1 + g_{be} r_B}{g_m + g_{be}} = \frac{r_B + \frac{1}{g_{be}}}{1 + \beta_0} \end{aligned}$$

$$\begin{aligned} \underline{Z}'_{21} &= \left. \frac{U'_2}{I'_1} \right|_{I'_2=0} : I'_1 = (g_{be} + g_m) \tilde{U}_{BE} ; U'_2 = U'_1 - R_F g_m \tilde{U}_{BE} = \tilde{U}_{BE} + r_B g_{be} \tilde{U}_{BE} - R_F g_m \tilde{U}_{BE} \\ \Rightarrow \underline{Z}'_{21} &= \frac{1 - r_b g_{be} - R_F g_m}{g_{be} + g_m} = \frac{\frac{1}{g_{be}} + r_B - \beta_0 R_F}{1 + \beta_0} \end{aligned}$$

$$\begin{aligned} \underline{Z}'_{22} &= \left. \frac{U'_2}{I'_2} \right|_{I'_1=0} : \tilde{U}_{BE} = U'_1 \frac{1}{1 + g_{be} r_B} \\ U'_1 &= U'_2 - R_F g_{be} \tilde{U}_{BE} \\ \Rightarrow \tilde{U}_{BE} &= U'_2 \frac{1}{1 + g_{be} r_B} - \frac{R_F g_{be}}{1 + g_{be} r_B} \tilde{U}_{BE} \Leftrightarrow (1 + g_{be} r_B + R_F g_{be}) \tilde{U}_{BE} = U'_2 \\ I'_2 &= (g_m + g_{be}) \tilde{U}_{BE} \\ \Rightarrow \underline{Z}'_{22} &= \frac{1 + g_{be} r_B + R_F g_{be}}{g_m + g_{be}} = \frac{\frac{1}{g_{be}} + r_B + R_F}{1 + \beta_0} \end{aligned}$$

$$\Rightarrow [\underline{Z}'] = \begin{pmatrix} \frac{1 + r_B g_{be}}{g_{be} + g_m} & \frac{1 + r_B g_{be}}{g_{be} + g_m} \\ \frac{1 + r_B g_{be} - R_F g_m}{g_{be} + g_m} & \frac{1 + r_B g_{be} + R_F g_m}{g_{be} + g_m} \end{pmatrix}$$

Rückkoppelzweitor:



$$\left. \begin{array}{l} \underline{Z''}_{11} = j\omega(L_1 + L_3) \\ \underline{Z''}_{12} = \underline{Z''}_{21} = j\omega L_1 \\ \underline{Z''}_{22} = j\omega(L_1 + L_2) \end{array} \right\} \Rightarrow [\underline{Z}''] = \begin{pmatrix} j\omega(L_1 + L_3) & j\omega L_1 \\ j\omega L_1 & j\omega(L_1 + L_2) \end{pmatrix}$$

$$[\underline{Z}] = [\underline{Z}'] + [\underline{Z}']$$

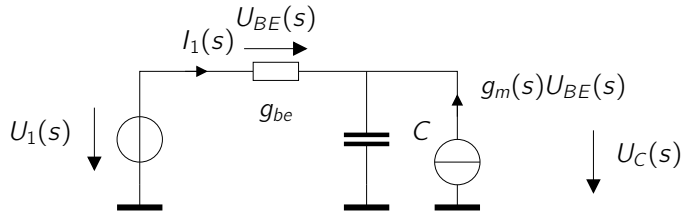
e) Verstärkung:

$$\begin{pmatrix} \underline{U}_1 \\ \underline{U}_2 \end{pmatrix} = \begin{pmatrix} \underline{Z}_{11} & \underline{Z}_{21} \\ \underline{Z}_{21} & \underline{Z}_{22} \end{pmatrix} \begin{pmatrix} \underline{I}_1 \\ \underline{I}_2 \end{pmatrix}$$

$$\Rightarrow \underline{V} = \left. \frac{\underline{U}_2}{\underline{I}_1} \right|_{\underline{I}_2=0} = \underline{Z}_{21}$$

## Aufgabe 2

a) Wechselstrom-Kleinsignal-Ersatzschaltbild:



b)

$$\begin{aligned}
 I_1(s) &= g_{be}U_{BE}(s) \\
 U_{BE}(s) &= U_1(s) - U_C(s) \\
 U_C(s) &= \frac{1}{sC}(I_1(s) + g_m(s)U_{BE}(s)) = \frac{1}{sC}(g_{be} + g_m(s))U_{BE}(s) \\
 \Leftrightarrow U_{BE}(s) &= U_1(s) - \frac{1}{sC}(g_{be} + g_m(s))U_{BE}(s) \Leftrightarrow U_{BE}(s) = \frac{1}{1 + \frac{1}{sC}(g_{be} + g_m(s))}U_1(s) \\
 \Rightarrow Y_1(s) &= \frac{g_{be}}{1 + \frac{1}{sC}(g_{be} + g_m(s))} = \frac{sCg_{be}}{sC + g_{be} + g_m(s)}
 \end{aligned}$$

c) Nein, da andere Ursache und damit andere Netzwerktopologie

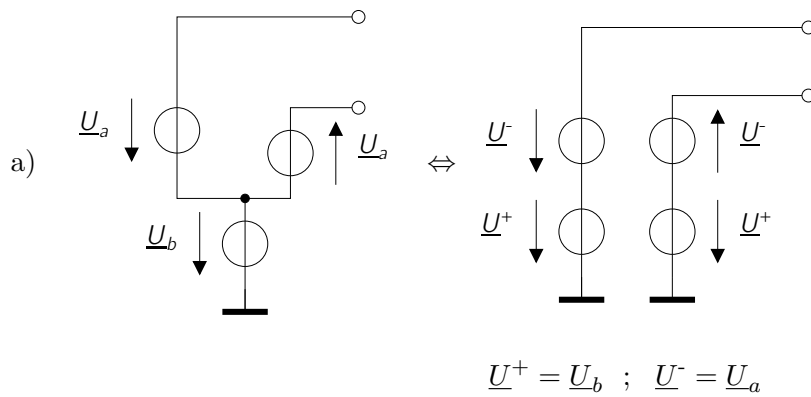
d) Polstelle, Wirkfunktion

$$\begin{aligned}
 Y_1(s) &= \frac{sCg_{be}}{sC + g_{be} + \frac{g_{m0}}{1 + \frac{s}{\omega_0}}} = \frac{sCg_{be}(1 + \frac{s}{\omega_0})}{sC(1 + \frac{s}{\omega_0}) + g_{be}(1 + \frac{s}{\omega_0}) + g_{m0}} \\
 sC(1 + \frac{s}{\omega_0}) + g_{be}(1 + \frac{s}{\omega_0}) + g_{m0} &= 0 \Leftrightarrow s^2 \frac{C}{\omega_0} + sC + s \frac{g_{be}}{\omega_0} + g_{be} + g_{m0} = 0 \\
 \Leftrightarrow s^2 + s(\omega_0 + \frac{g_{be}}{C}) + (g_{be} + g_{m0})\frac{\omega_0}{C} &= 0 \\
 \left(s + \frac{1}{2}(\omega_0 + \frac{g_{be}}{C})\right)^2 &= -(g_{be} + g_{m0})\frac{\omega_0}{C} + \left(\frac{1}{2}(\omega_0 + \frac{g_{be}}{C})\right)^2 \\
 \Leftrightarrow s_{1,2} &= -\frac{1}{2}(\omega_0 + \frac{g_{be}}{C}) \pm \sqrt{\frac{1}{4}(\omega_0 + \frac{g_{be}}{C})^2 - (g_{be} + g_{m0})\frac{\omega_0}{C}}
 \end{aligned}$$

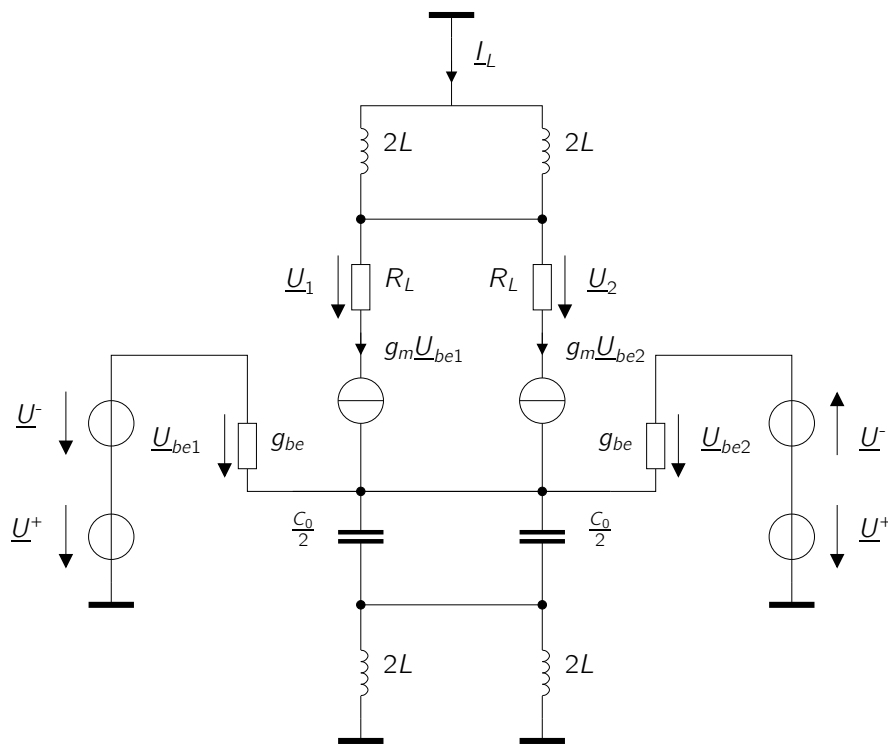
e) stabil  $\Rightarrow \operatorname{Re}\{s_{1,2}\} < 0$

$$\begin{aligned}
 &\Rightarrow \text{entweder Wurzel imaginär oder Wurzel} < +\frac{1}{2}(\omega_0 + \frac{g_{be}}{C}) \\
 &\Rightarrow (g_{be} + g_{m0})\frac{\omega_0}{C} > 0 \Rightarrow C < \infty \text{ und } C > 0
 \end{aligned}$$

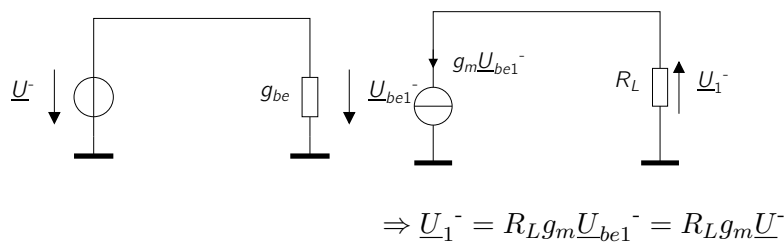
### Aufgabe 3



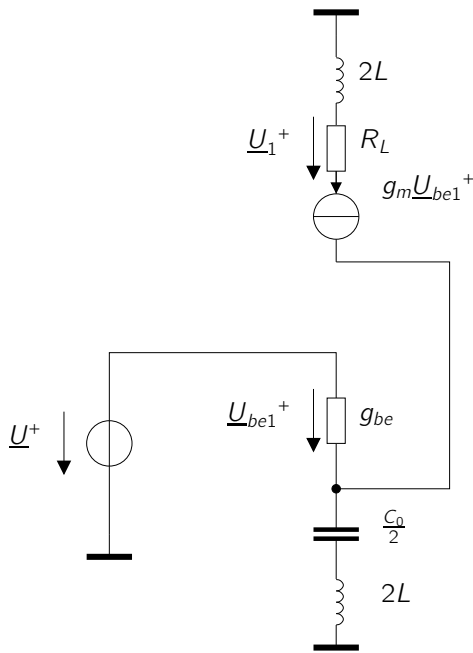
b) Ersatzschaltbild



Gegentakt:



Gleichtakt:



$$\begin{aligned} \underline{U}_1^+ &= R_L g_m \underline{U}_{be1}^+ \\ \underline{U}_{be1}^+ &= \underline{U}^+ - \left( j\omega 2L + \frac{1}{j\omega \frac{C_0}{2}} \right) (g_m + g_{be}) \underline{U}_{be1}^+ \\ \Leftrightarrow \underline{U}_{be1}^+ &= \frac{1}{1 + (g_{be} + g_m) \left( j\omega 2L + \frac{1}{j\omega \frac{C_0}{2}} \right)} \underline{U}^+ \\ \Rightarrow \underline{U}_1^+ &= \frac{g_m R_L}{1 + 2(g_{be} + g_m) \left( j\omega L + \frac{1}{j\omega C_0} \right)} \underline{U}^+ \end{aligned}$$

c) Spannung:

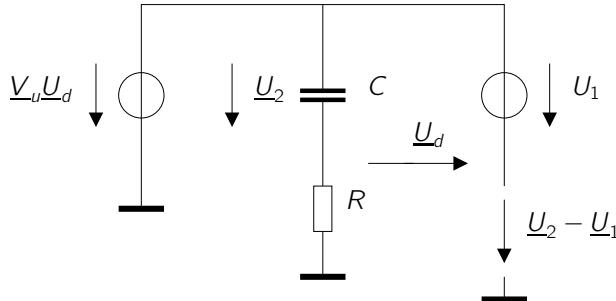
$$\underline{U}_1 = \underline{U}_1^+ + \underline{U}_1^- = \frac{g_m R_L}{1 + 2(g_{be} + g_m) \left( j\omega L + \frac{1}{j\omega C_0} \right)} \underline{U}_b + g_m R_L \underline{U}_a$$

d) siehe Originalschaltung :  $\underline{I}_2 = \frac{\underline{U}_1}{R_L} + \frac{\underline{U}_2}{R_L}$

$$\begin{aligned} U_1 &\text{ siehe (c)} \\ U_2 &= \underline{U}_2^+ + \underline{U}_2^- = \underline{U}_1^+ - \underline{U}_1^- \text{ (Symmetrie)} \\ \Rightarrow \underline{I}_2 &= \frac{\underline{U}_1^+ + \underline{U}_1^-}{R_L} + \frac{\underline{U}_1^+ - \underline{U}_1^-}{R_L} = 2 \frac{\underline{U}_1^+}{R_L} = \frac{2g_m}{1 + (g_{be} + g_m) \left( j\omega L + \frac{1}{j\omega C_0} \right)} \underline{U}_b \end{aligned}$$

## Aufgabe 4

a) Frequenzgang



$$\underline{U}_d = \frac{R}{R + \frac{1}{j\omega C}} \underline{U}_2 - (\underline{U}_2 - \underline{U}_1) = \left( \frac{j\omega RC}{1 + j\omega RC} - 1 \right) \underline{U}_2 + \underline{U}_1$$

$$= \underline{U}_1 - \frac{1}{1 + j\omega RC} \underline{U}_2$$

$$\underline{U}_2 = \underline{V}_u \underline{U}_D = \underline{V}_u \underline{U}_1 - \frac{\underline{V}_u}{1 + j\omega RC} \underline{U}_2 \Leftrightarrow \left( 1 + \frac{\underline{V}_u}{1 + j\omega RC} \right) \underline{U}_2 = \underline{V}_u \underline{U}_1$$

$$\Rightarrow \underline{F} = \frac{1}{\frac{1}{\underline{V}_u} + \frac{1}{1 + j\omega RC}}$$

b) Sonderfall:

$$\underline{F} \xrightarrow{\underline{V}_u \rightarrow \infty} 1 + j\omega RC$$

c)  $\underline{F}_2$ :

$$\underline{F} = \frac{1}{\frac{1}{\underline{F}_a} + \underline{F}_2} \xrightarrow{\underline{F}_a \rightarrow \infty} \frac{1}{\underline{F}_2} \stackrel{!}{=} 1 + j\omega RC \Leftrightarrow \underline{F}_2 = \frac{1}{1 + j\omega RC}$$

$$\frac{1}{\underline{F}_a} + \underline{F}_2 \stackrel{!}{=} \frac{1}{\underline{V}_u} + \frac{1}{1 + j\omega RC} \Leftrightarrow \underline{F}_a = \underline{V}_u$$

$$\underline{F}_0 = \underline{F}_2 \underline{F}_a = \frac{\underline{V}_u}{1 + j\omega RC}$$

d) Betrag und Phase der Schleifenverstärkung:

$$\underline{F}_0 = \frac{1}{a} \frac{V_0}{(1 + \frac{j\omega}{\omega_0})(1 + \frac{j\omega}{10\omega_0})(1 + \frac{j\omega}{100\omega_0})} \quad (\text{mit } a=1)$$

e)  $20 \text{ dB } \log\left(\frac{V_0}{a}\right) = 80 \text{ dB} \Leftrightarrow \frac{V_0}{a} = 10^4$

f) Die Schaltung ist instabil, da an der Stelle wo  $(\underline{E}_0)$  die 0 dB-Achse schneidet, die Phase  $\varphi(\underline{E}_0) < -180^\circ$  ist.

g) Schnittpunkt mit 0 dB-Achse muss bei  $w = 10 \omega_0$  sein.

$$\begin{aligned} \Rightarrow |\underline{E}_0(\omega = 0)| = 20 \text{ dB} &\Rightarrow 20 \text{ dB } \log\left(\frac{v_0}{a}\right) = 20 \text{ dB} \Rightarrow \frac{v_0}{a} = 10 \\ \Leftrightarrow v_0 = 10a = 10 \end{aligned}$$



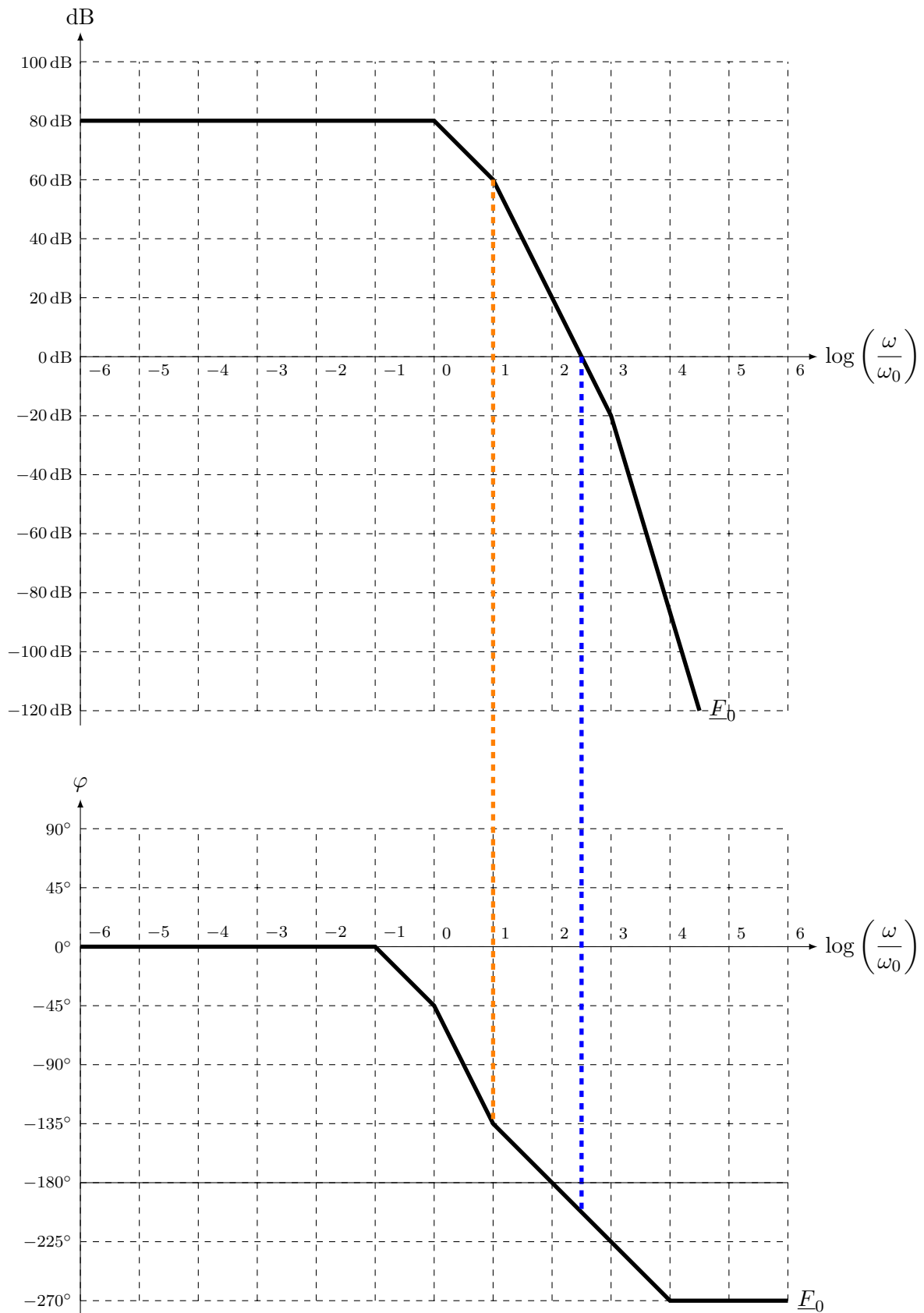


Abbildung 1: Lösung Bode-Diagramm.